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# 1 **Programmable Soliton Dynamics in All-Josephson-Junction Logic**

## 2 **Cells and Networks**

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## 14 **Abstract**

15 We demonstrate programmable control of kinetic soliton dynamics in all-Josephson-junction (all-  
16 JJ) networks through a novel tunable cell design. This cell enables on-demand switching of trans-  
17 mission lines and operates across defined parameter regimes supporting diverse dynamical modes.  
18 By introducing a structural asymmetry into a transmission line, we implement a Josephson diode  
19 that enforces unidirectional soliton propagation. The programmability of the kinetic inductance  
20 then provides a crucial mechanism to selectively enable or disable this diode functionality. By  
21 engineering artificial inhomogeneity into the circuit architecture, we enhance robustness in all-JJ  
22 logic circuits, 2D transmission line all-JJ lattices, and neuromorphic computing systems.

## Keywords

superconducting electronics; superconducting neural networks; kinetic inductance; soliton dynamics; Josephson-based diode

## Introduction

The rapid advancement of Josephson junction (JJ) logic circuits [1-5] and neuromorphic networks [6-9] holds transformative potential for ultra-low-power computing. However, achieving scalable integration remains a critical bottleneck, as conventional JJ-based architectures face fundamental density constraints imposed by magnetic flux manipulation requirements and complex mutual inductive crosstalks.

Circuits composed entirely of Josephson junctions (all-JJ circuits) [10-16] represent a promising platform for energy-efficient, high-speed and scalable computing. In these systems, the propagation of information is associated with the movement of a current wave / topological soliton, which is clearly visible in the model by a  $2\pi$ -jump of the so-called Josephson phase,  $\varphi$ . In contrast to conventional Rapid Single Flux Quantum (RSFQ) logic, the phase drop for the considered Single Kinetic Soliton (SKS) occurs not on the relatively large connecting geometric inductors, but on the Josephson junctions. SKS is a propagating wave of phase change with limited from below kinetic energy, the corresponding current pulse "dissipates" if its motion is interrupted, for example, by a structural inhomogeneity in a transmission line. Traditionally, this sensitivity to structural inhomogeneities has been viewed as a challenge for robust circuit design.

In this work, we propose to exploit the sensitivity mentioned above. We base our proposal on the concept of applying a small number of key cells, which should create precisely engineered tunable inhomogeneities. Such inhomogeneity may be designed as an element of tunable kinetic inductance [17]. This element has high inductance at small scales and can be controlled using currents [18,19], voltage [20] or magnetic fields [21,22]. At the same time, the use of hybrid superconductor-normal metal structures makes it possible to increase the effect of frequency tuning [23,24], while the addition of ferromagnetic layers permits the non-volatile control [25,26].

The another feature of tunable kinetic inductance element is the linear behaviour for weak signals, which excludes formation of parasitic processes in the transmission line. This permits to apply tunable kinetic inductance in the resonators with shifting resonance frequency [19,21,22], as well as in sensitive all-JJ digital circuits.

This idea enables us to use the 'flaws' of the structure as its important features, opening up a pathway to creating programmable and reconfigurable large circuits. An obvious and widely required application of this technology is in the development of superconductive programmable gate arrays (SPGA) [27-30], an active area of current research. Another important application of this idea lies in the promising neuromorphic direction [31-33]. Earlier in [34], we have already proposed using kinetic inductances to control neuron dynamics in networks based on radial basis functions (RBF-networks). Moreover, this approach can be extended to hardware realisations of bio-inspired spiking neural networks [35-42] by solving the challenges of creating controllable synapses to realize the effect of spike-timing-dependent plasticity and unidirectional feedbacks for self-regulation. Furthermore, the physical resemblance between solitons and the action potentials (spikes of voltage) generated in biological nervous systems makes all-JJ structures tempting candidates for constructing neuromorphic hardware [43].

In this paper, we investigate the use of controlled kinetic inductance to create an engineered inhomogeneous medium for kinetic solitons. We demonstrate that by tuning this inhomogeneity, distinct dynamical modes can be induced, fundamentally altering the soliton's behavior. Furthermore, we explore how structural asymmetry within this medium can be exploited to achieve a diode effect, enabling non-reciprocal soliton propagation. Building upon these foundational concepts, we then propose two specific architectural solutions: a programmable switch and a versatile routing matrix, which we term the "WayMatrix". We suggest that these architectures provide a framework for the flexible configuration of advanced logic and neuromorphic circuits.

## Results

### Model description

To model the dynamics of kinetic solitons [43], we employ the Resistively and Capacitively Shunted Junction (RCSJ) model [1], where the total current  $I$  across a Josephson junction is the sum of the supercurrent, the quasi-particle current, and the displacement current:

$$I = I_c \sin(\varphi) + \frac{V}{R_N} + C \frac{dV}{dt}. \quad (1)$$

Here,  $\varphi$  is the phase difference for the complex superconducting order parameter across the junction,  $V$  is the voltage,  $I_c$  is the critical current,  $R_N$  is the resistance in normal state and  $C$  is the capacitance. For analysis, it is convenient to express this equation in a dimensionless form. We normalize the time to the inverse of a reference plasma frequency,  $\tau = \tilde{\omega}_p t$ , where  $\tilde{\omega}_p = \sqrt{2\pi\tilde{I}_c/(\Phi_0\tilde{C})}$ , and normalise the current to a reference critical current  $\tilde{I}_c$ . This yields:

$$i = A \cdot \sin(\varphi) + \alpha \dot{\varphi} + \ddot{\varphi}. \quad (2)$$

In this normalized equation, the dots above the phases indicate differentiation over time with respect to  $\tau$ . The dimensionless damping coefficient is  $\alpha = \Phi_0\tilde{\omega}_p/(2\pi\tilde{I}_cR_N)$ . The term  $\dot{\varphi}$  represents the voltage normalized by the characteristic voltage  $V_0 = \Phi_0\tilde{\omega}_p/(2\pi)$ . The parameter  $A = I_c/\tilde{I}_c$  is the normalized amplitude of the critical current for junctions with the critical current  $I_c$  that differs from the reference normalization value  $\tilde{I}_c$ .

To analyze the circuit dynamics, we adopt a nodal analysis approach. In this approach, the gauge-invariant phase difference across any element is expressed in terms of the nodal phases at its terminals,  $\varphi = \varphi_k - \varphi_j$ . The phase of the ground node is set to zero by convention. This formulation inherently satisfies Kirchhoff's Current Law (KCL) at each node. For any node  $k$  connected to  $H$

elements, KCL dictates that the algebraic sum of currents is zero:

$$\sum_{h=1}^H I_{k,j(h)} = 0, \quad (3)$$

where the index  $h$  runs over all elements connected to node  $k$ ,  $j(h)$  is the index of the node at the other end of element  $h$ , and  $I_{k,j(h)}$  is the normalised current flowing from the node  $k$  to the node  $j(h)$ . Each current is described by the RCSJ model (Eq. 2):

$$I_{k,j(h)}/\tilde{I}_c = A_h \sin(\varphi_k - \varphi_{j(h)}) + \alpha_h(\dot{\varphi}_k - \dot{\varphi}_{j(h)}) + (\ddot{\varphi}_k - \ddot{\varphi}_{j(h)}). \quad (4)$$

With this approach, the current across the inductance is defined by the expression

$$I_{k,j(h)}^L/\tilde{I}_c = \frac{\Phi_0}{2\pi L\tilde{I}_c}(\varphi_k - \varphi_{j(h)}) = (\varphi_k - \varphi_{j(h)})/l, \quad (5)$$

where  $l = L/L_J$  is inductance normalised to the Josephson inductance  $L_J = \Phi_0/(2\pi\tilde{I}_c)$ .

After substituting the expressions for the current into the formula for the current balance at the node, we get:

$$M_{k,k}\ddot{\varphi}_k - \sum_{h=1}^H M_{k,j(h)}\ddot{\varphi}_{k,j(h)} = F_k(\varphi_k, \varphi_{k,j(h)}, \dot{\varphi}_k, \dot{\varphi}_{k,j(h)}), \quad (6)$$

where  $M_{k,k}$  is the sum of the coefficients before  $\ddot{\varphi}_k$ ,  $M_{k,j(h)}$  are the coefficients before  $\ddot{\varphi}_{k,j(h)}$ ,  $F_k$  contains the sum of all summands except those that do not contain the second derivative. In  $F_k$ , all summands with  $\varphi_k$  are written with a minus sign, and all summands with  $\varphi_{k,j(h)}$  are written with a plus sign. Additional currents (e.g. the bias current or the time-dependent current from the generator) are also included as components. After writing down the equations 6 for each node, a system of second-order diffeomorphic equations are obtained, which can be represented in matrix

112 form:

$$113 \quad \hat{M}\varphi = \begin{pmatrix} M_{11} & M_{12} & \dots & M_{1N} \\ M_{21} & M_{22} & \dots & M_{2N} \\ \vdots & \vdots & \ddots & \vdots \\ M_{N1} & M_{N2} & \dots & M_{NN} \end{pmatrix} \begin{pmatrix} \ddot{\varphi}_1 \\ \ddot{\varphi}_2 \\ \vdots \\ \ddot{\varphi}_N \end{pmatrix} = \mathbf{F}(\varphi, \varphi). \quad (7)$$

114 The resulting system of  $N$  ordinary differential equations is expressed in the matrix form shown in  
 115 Eq. 7. In this equation,  $\vec{\varphi}$  is the vector of nodal phases,  $N$  is the total number of non-ground nodes,  
 116 and  $\hat{M}$  is the  $N \times N$  mass matrix (also known as the capacitance matrix), which is defined by the  
 117 capacitive coupling coefficients from Eq. 6. A key property of  $\hat{M}$  is its sparsity, which arises di-  
 118 rectly from the local connectivity of the circuit topology; each node is connected to a small sub-  
 119 set of other nodes. To increase computational efficiency, we exploit this sparsity when solving the  
 120 system. The equations are integrated numerically using an adaptive-step-size solver based on the  
 121 explicit Runge-Kutta ( $4^{th}$  and  $5^{th}$  order) formula, commonly known as the Dormand-Prince pair  
 122 [44,45], which is well-suited for this class of non-stiff problems.

## 123 The Kinetic Inductance Controllable Key

124 The fundamental building block of our design is the Kinetic Inductance Controllable Key (KICK),  
 125 which is constructed from the two modified unit cells of an all-Josephson Junction Transmission  
 126 Line (all-JJTL). As depicted in Figure 1a, each cell is modified by incorporating a controlled ki-  
 127 netic inductance in series with one of its Josephson junctions connected to the ground plane. There  
 128 are some operational regimes inherent to such KICK governed by the value of this inductance and  
 129 by the damping parameter of junctions within the transmission line. The damping parameter is a  
 130 critical factor as it dictates the kinetic soliton's propagation rate.

131 As a preliminary step, we characterized the dependence of the kinetic soliton propagation velocity  
 132 on the damping parameter of the connecting junctions,  $\alpha$  (see 1b). We define the velocity as the  
 133 number of grounded junctions traversed per unit of normalized time,  $\tau$ . Our simulations revealed

134 a critical damping threshold at  $\alpha_{crit} \approx 0.8$ ; below this value stable soliton propagation is not sup-  
 135 ported. Besides, under this condition the energy dissipation rate is too high relative to the energy  
 136 transfer between adjacent junctions, causing the soliton to decay. In case when  $\alpha > \alpha_{crit}$ , the soli-  
 137 ton velocity is a monotonically increasing function of the damping. This dependence falls into an  
 138 approximately linear regime for  $\alpha > 3$ . The physical mechanism for this velocity increase can be  
 139 understood from the RCSJ model: a higher value of  $\alpha$  enhances the resistive quasiparticle current  
 140 ( $\alpha\dot{\varphi}$ ) that flows as a junction switches. This larger current provides a stronger driving force to the  
 141 next junction in the line, causing it to reach its critical threshold and switch more rapidly, thus in-  
 142 creasing the overall propagation velocity of the soliton.

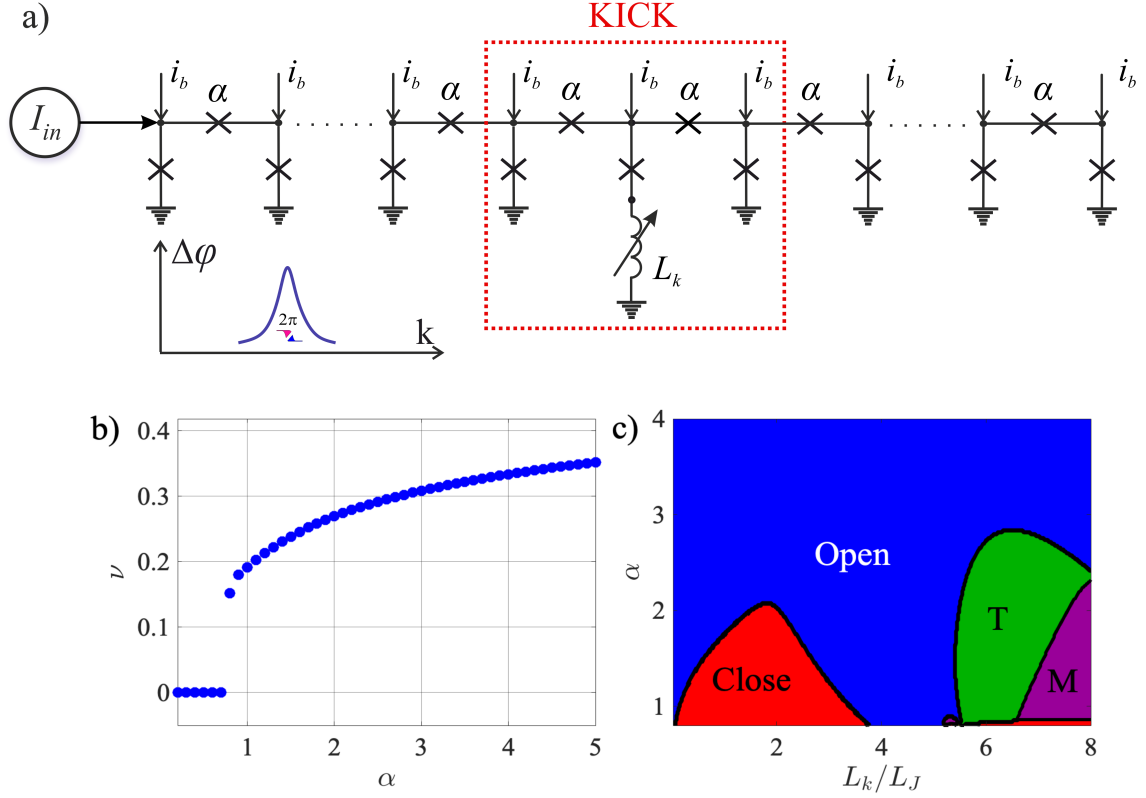
143 The functionality of the KICK is determined by the interplay between the damping  $\alpha$  and the nor-  
 144 malized kinetic inductance  $L/L_J$ . Figure 1c summarizes the behavior of the device in a parameter  
 145 map, which reveals four distinct operational regimes.

- 146 • **Open Mode:** The KICK is effectively transparent, allowing an incident kinetic soliton to  
 147 propagate through it with minimal perturbation.
- 148 • **Close Mode:** The KICK acts as a terminator, blocking and destroying the incoming soliton.
- 149 • **T-Mode:** The KICK functions as a T-flip-flop. It possesses two stable states, and each arriv-  
 150 ing soliton toggles the cell from its current state to the other. Every second soliton passes to  
 151 the exit.
- 152 • **M-Mode (Multy-State Mode):** This regime is characterized by the formation of more than  
 153 two stable states and other complex dynamics, which fall outside the scope of this study.

154 An essential feature of the KICK is the ability to switch between different modes at a fixed alpha  
 155 value: thus, by fixing alpha (for example,  $\alpha = 2$ ) and varying the kinetic inductance, we can switch  
 156 between all modes (**Open Mode**  $\rightarrow$  **Close Mode**  $\rightarrow$  **Open Mode**  $\rightarrow$  **T-Mode**  $\rightarrow$  **M-Mode**)  
 157 represented on the parameter map (see Figure 1).

158 To illustrate the operational modes of the KICK, we simulated the propagation of a kinetic soliton  
 159 through the all-JJTL. The simulated line comprises 31 grounded junctions with a uniform damping





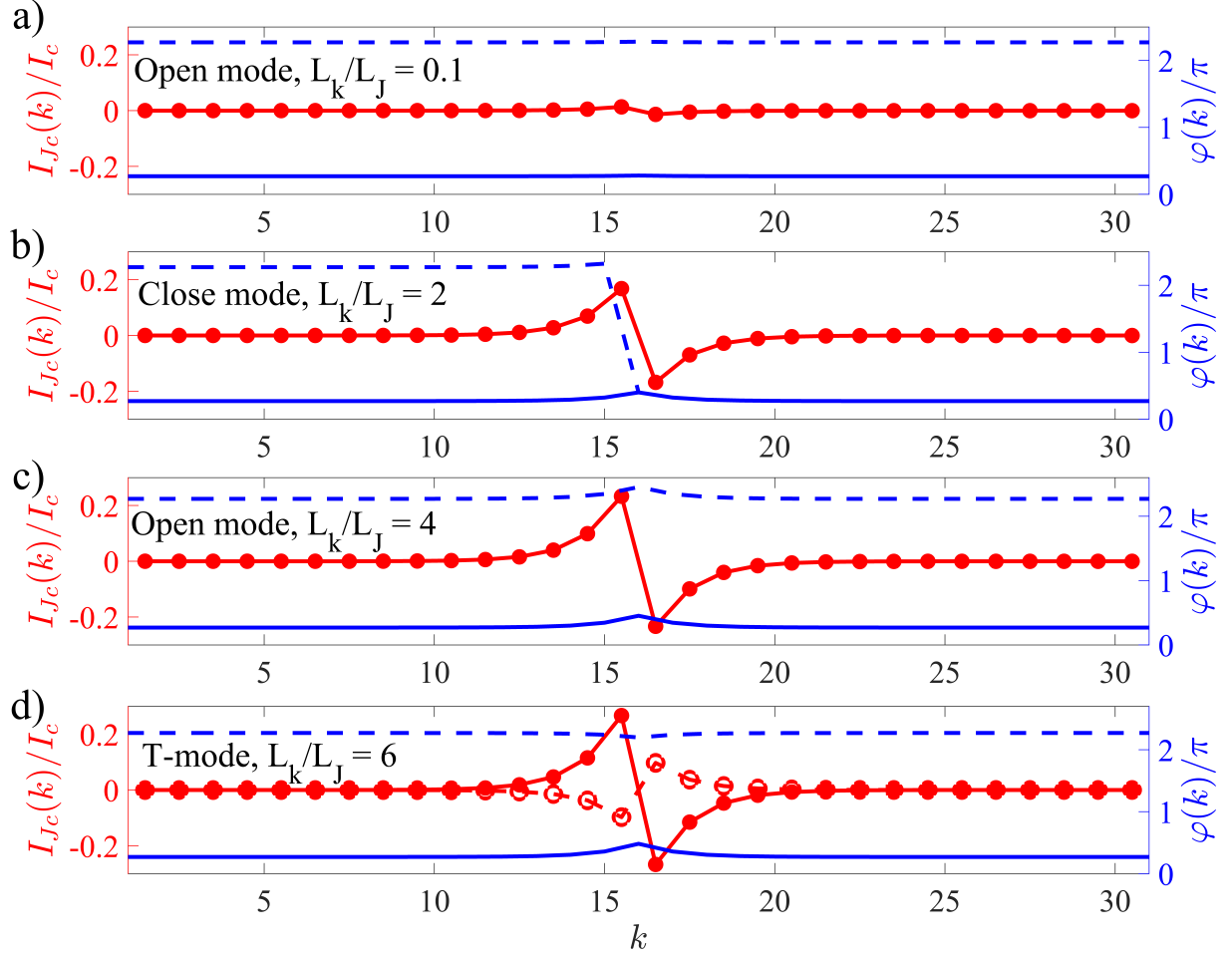
**Figure 1:** **a)** An equivalent scheme for the Kinetic Inductance Controllable Key (KICK) as a part of an all-Josephson transmission line. A soliton which dynamics is controlled by the key developed is represented schematically. **b)** Dependence of the single kinetic soliton (SKS) propagation velocity, measured in Josephson junctions per normalized time unit, on the damping parameter  $\alpha$ ; **c)** Map of different modes depending on the damping parameter and kinetic inductance measurement. Close mode (red zone): the KICK does not allow SKS to pass through. Open mode (blue zone): the KICK passes SKS. T-mode (green zone): the KICK has two stable states and every second SKS passes through it. M-mode (purple zone): the KICK has many stable states.

parameter of connecting junctions  $\alpha = 1$ . The KICK is implemented by inserting a controlled kinetic inductance in series with the ground junction at the line's center (node  $k = 16$ ). Figure 2 presents the results for different values of this inductance, corresponding to distinct operational modes. Each panel displays two key physical quantities on dual y-axes:

- 1) The spatial profile of the nodal Josephson phases ( $\varphi_k$ ) as a function of the node index  $k$ .
- 2) The normalized currents flowing through the series junctions connecting the nodes. The current between nodes  $k$  and  $k + 1$  ( $I_c/\tilde{I}_c \sin(\varphi_{k+1} - \varphi_k)$ ) is plotted at the midpoint index  $k + 0.5$  for visual clarity. This visualization allows for a direct comparison of the system's state before and after soliton interaction. The solid lines depict the initial state (before the soliton reaches the KICK), and the dashed lines show the final state (after the soliton has passed and the system has settled).

For a low inductance of  $L/L_J = 0.1$  (see Figure 2a), corresponding to the Open Mode, the KICK causes only slight disturbance in the transmission line. The incident soliton propagates through it unimpeded, and the entire line returns to its initial physical state. However, increasing the inductance to  $L/L_J = 2$  (see Figure 2b) switches the system to the Close Mode. In this mode, the KICK serve as a significant barrier; when the soliton arrives, the large inductance impedes the necessary current dynamics, halting the propagation and causing the soliton to be annihilated. Consequently, the  $2\pi$  phase slip, which signifies the soliton's passage, only traverses the first half of the line (nodes 1 to 15), while the segment beyond the KICK remains entirely unperturbed. Remarkably, a further increase of inductance to  $L/L_J = 4$  (see Figure 2c) leads to the re-emergence of the Open Mode. This non-trivial effect is governed by transient energy storage in the inductor  $L$ . Although the soliton is momentarily halted at the KICK, the subsequent release of stored magnetic energy provides the necessary "kick" to complete the phase slip at node 16. This re-initiates the propagation, allowing the soliton to effectively re-form and travel down the rest of the line. Similarly to the low inductance case, the soliton successfully traverses the entire line, and the system returns to its initial physical state.

The behavior of the KICK in the T-Mode, which enables its use as a T-flip-flop, is detailed in Figure 2d. This mode is defined by the existence of two distinct stable states, physically correspond-



**Figure 2:** Spatial distributions of the Josephson phase (blue curves) and current (red curves) in: (a,c) open mode,  $L_k/L_J = 0.1$  (a) and  $L_k/L_J = 4$  (c); (b) close mode,  $L_k/L_J = 2$ ; and (d) T-mode,  $L_k/L_J = 6$ . Solid lines show initial profiles, dashed lines represent distributions after soliton passage. The Josephson phase is plotted against the integer node index  $k$ , whereas the current is plotted at the midpoint index  $k + 0.5$  to represent the junction between nodes  $k$  and  $k + 1$ , see Figure 1a.

187 ing to a bistable potential landscape created by the KICK architecture. These two states are distin-  
188 guished by the presence of persistent, static currents of opposite polarity flowing from the central  
189 node ( $k = 16$ ). This physical difference leads to an fundamentally state-dependent and asymmetric  
190 toggling action. When the KICK is in the first stable state, an incoming soliton successfully flips it  
191 to the second state and is transmitted, continuing its propagation down the line. Conversely, when  
192 starting from the second state, an arriving soliton again flips the KICK back to the first state, but is  
193 annihilated in the process and does not propagate further. This state-dependent transmission and  
194 annihilation is the core mechanism that allows the KICK to function as a memory element or a dy-  
195 namic routing switch.

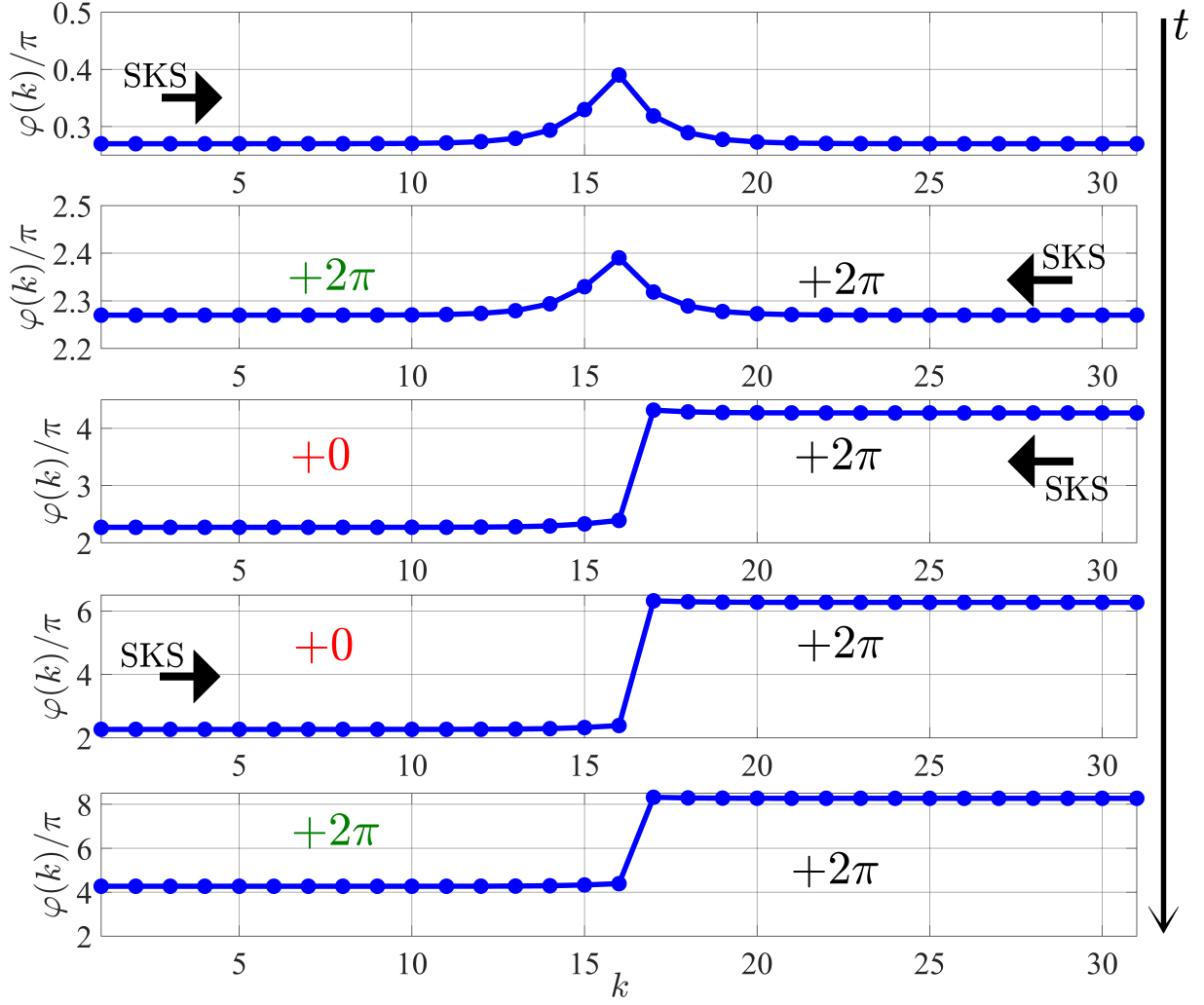
196 Beyond primary operational modes, the system exhibits other notable behavior types in specific  
197 regions of its parameter space. The M-Mode, for instance, is characterized by complex responses,  
198 depend on previous events. This can include such behavior when an initial soliton is annihilated,  
199 effectively "priming" the cell to transmit all subsequent solitons, a feature potentially useful for  
200 tasks like sequential filtering. Furthermore, in the transition regions between the primary modes,  
201 we observe phenomena such as soliton reflection back towards the source.

202 Finally, the asymptotic behavior in the high-damping ( $\alpha$ ) limit is particularly significant. As  $\alpha$  in-  
203 creases, so does the soliton's velocity and kinetic energy. Consequently, for sufficiently high  $\alpha$ , the  
204 soliton's energy is large enough to overcome any potential barrier presented by the KICK, ensur-  
205 ing transmission regardless of the inductance value. This results in a universal Open Mode at high  
206 rates. Crucially, this high-energy passage is not inert; if the KICK is in a bistable regime (such as  
207 the T-Mode), the "passing" soliton can still deliver enough of an impulse to toggle the cell's state.

## 208 **The Soliton Diode**

209 What is even more interesting is that the KICK architecture can be engineered to function as a soli-  
210 ton diode, a device the function of which is similar to that of a semiconductor diode, allowing the  
211 soliton to pass in only one direction. This is achieved by introducing a structural asymmetry into  
212 the cell's design. It is important to note that such non-reciprocal behavior can be achieved even

without the kinetic inductance ( $L = 0$ ). However, the inclusion of one (a tunable inductance) is a  
 key innovation, as it allows to dynamically switch this directional property on and off.  
 We demonstrate this principle through simulation of a KICK with  $L/L_J = 2$ . In our model, the  
 transmission line's series junctions have a nominal critical current of  $I_c = 0.7\tilde{I}_c$ . The asymmetry  
 is created by increasing the critical current of the specific junction connecting nodes 15 and 16 to  
 $I_c = \tilde{I}_c$  (i.e., to 1 in normalized units). The effect of this asymmetric potential barrier is that a  
 soliton initiated in the forward direction (from node 1) successfully overcomes it and is transmitted  
 along the entire line. In contrast, a soliton propagating in the reverse direction (from node 31) is  
 unable to pass the barrier and is annihilated at node 17.  
 Figure 3 demonstrates the non-reciprocal behavior of the soliton diode by showing a sequence  
 of five snapshots of the nodal Josephson phase distribution at successive moments in time, ar-  
 ranged from top to bottom. The process begins with the line in its initial state (top panel), after  
 which a soliton is initiated from the left side (node 1). As shown in the second panel, this forward-  
 propagating soliton successfully passes through the diode, resulting in a  $2\pi$  phase advance across  
 all nodes. Immediately after, a new soliton is initiated from the right side (node 31) to test the  
 reverse direction. The third panel reveals that this soliton is blocked; its propagation is halted at  
 the diode, and the corresponding  $2\pi$  phase slip is confined to nodes 17 through 31. The fourth  
 panel confirms the robustness of this blocking action, as a second, subsequent reverse-propagating  
 soliton is also annihilated in the same manner. To complete the demonstration, another forward-  
 propagating soliton is sent from the left. The fifth panel confirms that the diode once again allows  
 it to pass, resulting in another full  $2\pi$  phase advance across the entire line. It is crucial to note that  
 although the absolute phase values accumulate in multiples of  $2\pi$  throughout this sequence, the  
 physical state of the structure remains unchanged after each full transmission, a direct consequence  
 of the  $2\pi$  periodicity of the Josephson energy.  
 A significant feature of this structure is the ability to disable the diode effect. By increasing the  
 inductance to  $L/L_J = 3$ , the device becomes bi-directionally transparent, effectively turning the  
 diode function "off". This demonstrates how the introduced structural asymmetry alters the oper-



**Figure 3:** Temporal evolution of Josephson phase asymmetry in a soliton diode: (a) Initial state; (b) After left-propagating soliton passage; (c,d) Sequential right-propagating soliton interactions; (e) Final left-propagating soliton recurrence.

240 ational landscape of the device: an inductance value that would normally correspond to the Close  
241 Mode in a symmetric KICK now matches to a bi-directional Open Mode for the asymmetric diode  
242 structure. Furthermore, it is worth noting that the the asymmetry required for diode-like behavior  
243 can be achieved through alternative means, such as by creating a local mismatch in the damping  
244 parameter — for instance, by increasing  $\alpha$  from 1 to 3 for one of the series junctions instead of the  
245 critical current.

246 The ability to enforce a specific direction of soliton flow makes the soliton diode an essential com-  
247 ponent for complex circuit design. This is particularly critical in architectures involving feedback  
248 loops, where it is necessary to unambiguously define the direction of signal propagation. This con-  
249 cept can be extended by cascading two such tunable diodes with opposing forward directions. This  
250 configuration creates a programmable transmission line where the permitted direction of soliton  
251 travel can be pre-configured by setting the inductance values of each diode.

## 252 **Discussion**

### 253 **Implementation of reconfigurable networks**

254 On the basis of the operational principles of the Kinetic Inductance Controllable Key and the soli-  
255 ton diode, we now demonstrate how these fundamental building blocks can be integrated to create  
256 reconfigurable soliton-based logic circuits. We begin by proposing a specific proof-of-concept de-  
257 sign for a signal routing network and then introduce a generalized, scalable architecture suitable for  
258 complex computational tasks.

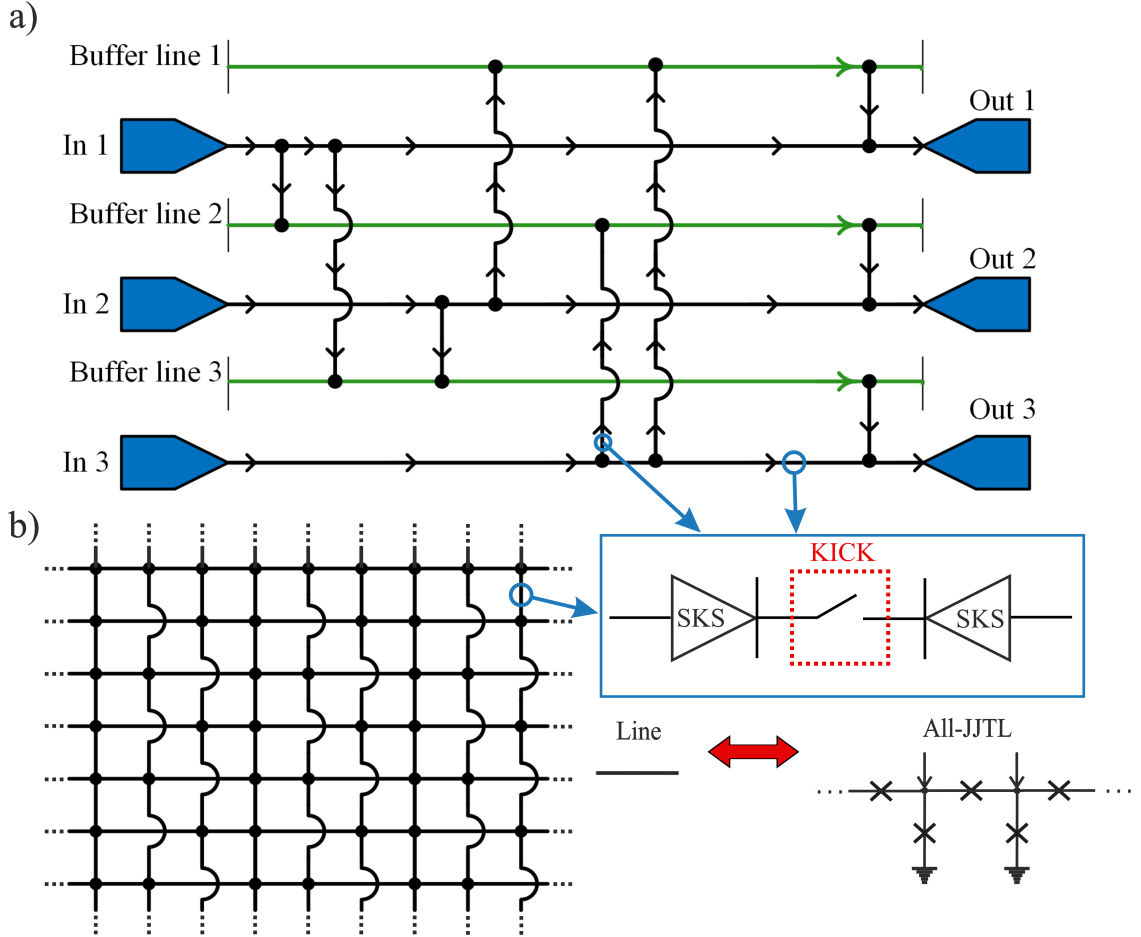
259 As a direct application of the KICK's switching capabilities, we first propose the 3-input, 3-output  
260 routing network illustrated in Figure 3a. The proposed architecture is based on a grid where each  
261 path depicted is itself a complete all-Josephson-junction transmission line (all-JJTL). The routing  
262 mechanism would depend on the incorporation of KICKs into specific segments of these all-JJTLs.  
263 By programming each KICK to be in either its Open Mode (transmitting) or Close Mode (block-  
264 ing), one could control the flow of solitons through the network and define a unique path from  
265 any input to any output. To prevent collisions between solitons traveling along different routes,

266 the design incorporates auxiliary buffer lines. These lines make it possible to define a set of non-  
 267 intersecting paths for all required connections, thus ensuring collision-free operation. This design  
 268 serves to validate the fundamental principle of using KICKs as programmable switches.

269 With this idea, we propose a more general and powerful architecture, which we term the "Way-  
 270 Matrix", shown schematically in Figure 3b. This versatile  $N \times M$  routing matrix is conceived as a  
 271 core component of larger soliton-based processors. Its enhanced functionality would be predicated  
 272 on the synergistic action of its core components. Firstly, KICKs integrated into the line segments  
 273 would act as programmable switches controlling the signal flow. Secondly, the directionality of  
 274 soliton propagation would be rigorously enforced by integrated soliton diodes. Thus, the diodes  
 275 and switches placed in the all-JJTL lines determine the direction of soliton propagation in the line.  
 276 Finally, to solve the problem of collisions in a dense matrix, we propose dedicated vertical lines  
 277 that enable row-skipping connections. For the same purposes, horizontal lines can also be used for  
 278 column-skipping connections.

279 At first glance, it may seem that the proposed architecture is a complicated version of a memris-  
 280 tive crossbar, but this is not the case. The main distinction is in the organization of interconnections  
 281 between lines: in a memristive crossbar, as the name follows, these connections are formed by the  
 282 intersection of signal lines and the corresponding memristive layer. In the proposed WayMatrix,  
 283 however, the lines are combined into a single node at the intersection point, the current direction  
 284 of which can be controlled by switches and diodes. The power of the WayMatrix architecture lies  
 285 in its potential use as a universal framework for creating programmable and reconfigurable con-  
 286 nections between different circuit blocks. WayMatrix makes it easy to set up feedback loops be-  
 287 tween these blocks, change their connection order, and perform logical operations. We envision it  
 288 serving as a reconfigurable "backbone" to link various specialized functional units within a larger  
 289 integrated circuit. For example, the WayMatrix could be configured to connect arrays of memory  
 290 cells to arithmetic logic units or to route data between different processing cores. Another key ap-  
 291 plication is the creation of programmable clock distribution networks. In such a role, the WayMa-  
 292 trix could manage signal timing across a chip by introducing precise, configurable delays into the





**Figure 4:** a) Schematic of an all-Josephson-junction transmission line (all-JJTL) network with three inputs (In 1, In 2, In 3), three outputs (Out 1, Out 2, Out 3), and three auxiliary buffer lines. Black arrows on the lines indicate soliton propagation paths. Input-output connections are configured by setting operation modes of kinetic inductance controllable keys (KICKs), where each cell either transmits or blocks solitons based on its programmed state. b) The schematic shows a transmission line matrix where path selection is governed by KICKs and signal directionality is ensured by soliton diodes. Specific vertical lines enable row-skipping connections to prevent soliton collisions during signal propagation.

293 clock paths, which is crucial for asynchronous circuit design. This would allow a single hardware  
294 platform to be flexibly repurposed for different algorithms by simply re-programming the routing  
295 paths, a paradigm central to the development of superconducting programmable gate arrays (SP-  
296 GAs).

297 The true potential of this architecture, however, is most evident in its application as an axon-  
298 synaptic connection matrix for neuromorphic computing. The ability to program connections, en-  
299 force directionality, and reconfigure paths makes the WayMatrix an ideal candidate for emulating  
300 the complex and plastic connectivity of a biological neural network. In such a system, each soliton  
301 acts as a "spike", and the WayMatrix serves as the synaptic network that routes these spikes be-  
302 tween artificial neurons. This lays the groundwork for building powerful, event-driven and energy-  
303 efficient spiking neural networks based on the principles we have outlined. In addition to using the  
304 WayMatrix, we can reconfigure the neural network itself, program connections between different  
305 neurons, implement synaptic pruning, and even "kill" part of the artificial brain.

306 Human or animal brain contains a huge number of synapses, many times greater than the num-  
307 ber of neurons (e.g. the Norwegian rat brain contains about 200 million neurons, each of which  
308 roughly has an average of about 1000 synapses [46]). The ability of a living being to solve certain  
309 tasks depends precisely on the number of inter-neuronal connections. In their attempts to imple-  
310 ment such complex systems in hardware, engineers and scientists inevitably face the problem of  
311 interconnects and the implementation of a huge number of synaptic connections. The supercon-  
312 ducting axon-synaptic matrix based on the WayMatrix concept seems to be a promising solution to  
313 the problem [47-50].

314 As mentioned above repeatedly, the field applications of kinetic inductance and, in particular,  
315 KICK, are also extended to bio-inspired neuromorphic spiking networks. One important feature  
316 of living nervous tissues is the ability to modulate the synaptic delay of signal propagation from  
317 one neuron to another. This feature is equally important to implement in hardware artificial real-  
318 isations of neuromorphic networks. The signal propagation delay is also affected by a length and  
319 a conductivity of an axon, which is quite simply imitated by means of a standard Josephson trans-

mission line, as well as by means of all-JJTL, discussed at the beginning of this article. A simple solution to modulate the propagation delay is to change the length (number of JTL cells) of such an artificial axon, but there is another way. The inductance connected in parallel with the Josephson junctions determines the amount of magnetic energy stored within each JTL cell. Consequently, a larger inductance value results in a longer propagation delay.

## Conclusions

This study demonstrates the programmable control of kinetic soliton dynamics in all-Josephson-junction networks through a novel tunable element, the Kinetic Inductance Controllable Key (KICK). By engineering inhomogeneity via controlled kinetic inductance, we induce distinct dynamical modes (Open, Close, T-Mode) that fundamentally alter soliton propagation. Furthermore, the features of the proposed cell enable a soliton diode effect, achieving non-reciprocal signal transmission. Building on these principles, we propose two scalable architectures: a programmable switch for reconfigurable routing and the WayMatrix, a versatile  $N * M$  routing matrix. These solutions establish a framework for robust, high-speed superconducting logic that addresses critical bottlenecks in this type of computing.

We realize that the time required to "reprogram" kinetic inductance significantly exceeds the picosecond timescales of Josephson junction dynamics. However, this re-configuration time should be considered in the context of hardware development cycles. From this point of view, the re-configuration time is orders of magnitude lower than the time required to design, fabricate, and test a new application-specific integrated circuit (ASIC), offering a compelling advantage in flexibility and prototyping rate.

The superconducting diodes proposed in this work can be used as a part of synaptic connections in neuromorphic networks to prevent the backward influence of a postsynaptic neuron on a presynaptic neuron through the same connection link. It should also be noted that the signal propagation time between neurons can be controlled by modulating the bias currents, the value of which directly affects the potential barrier in Josephson line (standard JTL or All-JJTL). Thus, the choice of

a particular method of signal propagation delay influence depends on the realization of interneuron interactions and the need to adjust a particular interneuron connection. Moreover, these approaches can be combined into one by using a chain of superconductor diodes. Using cells with kinetic inductances, we can change the local propagation speed of spikes in inter-neuronal signal transmission circuits by smoothly adjusting the delay time. The integration of the WayMatrix will make it possible to change the length of the axonal line as a whole, and thus introduce a delay. Besides, it is really interesting to examine how the dynamics of voltage spike formation in a bio-inspired neuron, proposed in [42], will change if we substitute geometric inductances for kinetic ones. Further development of the idea presented in this article will also address this aspect.

The proposed technique allows for a more compact design and new (diode) functionality of various superconducting computing modules, and makes possible further increase of integration density compared to well-known RSFQ technology.

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